

# Stability analysis and stabilization of semi-Markov jump linear systems with unavailable sojourn-time information

Xiaotai WU<sup>1</sup>, Yang TANG<sup>2\*</sup>, Shuai MAO<sup>2</sup> & Ying ZHAO<sup>1</sup><sup>1</sup>*School of Mathematics and Physics, Anhui Polytechnic University, Wuhu 241000, China;*<sup>2</sup>*Key Laboratory of Advanced Control and Optimization for Chemical Processes, Ministry of Education, East China University of Science and Technology, Shanghai 200237, China*

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**Abstract** This study addresses the stability and stabilization problems of discrete-time semi-Markov jump linear systems (S-MJLSs) with unavailable sojourn-time information. The sojourn-time probability mass functions (S-TPMFs) of discrete-time semi-Markov chains are no longer confined to the geometric distribution, and it is difficult to obtain accurate and comprehensive information for S-TPMFs in practice. This is because S-TPMFs are usually deduced from the statistical characteristics according to the sampled-data, while adequate samples are often costly and time consuming. In this study, when the S-TPMFs for semi-Markov chains are assumed to be unavailable, the  $\sigma$ -error mean square stability is investigated for discrete-time S-MJLSs with some widely used assumptions; for semi-Markov chains, only the transition probability matrix of the embedded chain is used. In addition, the existence conditions of the effective controller are provided for closed-loop systems without using the information of S-TPMFs. Numerical examples are presented to illustrate the validity of the obtained theoretical results.

**Keywords** semi-Markov chain, semi-Markov jump linear systems,  $\sigma$ -error mean square stability, unavailable sojourn-time information

## 1 Introduction

Stochastic switching commonly exists in many realistic systems, such as aircraft [1], electric circuit systems [2], vehicles [3], and communication networks [4], which are usually caused by external and internal disturbances, for example, configuration changes, unexpected events, and random faults [5–7]. Markov jump systems, as an important part of stochastic switching systems, have been widely investigated, for linear systems [5, 6, 8–12] and nonlinear systems [13–19]. Markov chains have memoryless property, which demands that the sojourn-time between consecutive mode switching should satisfy the geometrical distribution for discrete-time case, and exponential distribution for continuous-time case [20], respectively. This restricts the application scope of Markov jump systems.

The semi-Markov jump system is an extension of the Markov jump system, whose sojourn-time is unnecessarily confined to either geometric distribution or exponential distribution. Consequently, semi-Markov jump systems are capable of depicting a wider range of stochastic switching systems [21–25]. At the same time, the generality of the semi-Markov chain also brings more challenges to the stability analysis for semi-Markov jump systems. Numerous results have been reported on the stability and stabilization problems of such systems [26–38].

Recent years have shown that the semi-Markov kernel approach is effective in tackling the stability and stabilization problems for discrete-time S-MJLSs [27, 28, 39–43]. For example, the  $m$ th mean stability is presented for discrete-time and continuous-time positive S-MJLSs with state resets in [27]. The  $\sigma$ -error mean square stability ( $\sigma$ -MSS<sup>1</sup>) has been widely used to study S-MJLSs, (e.g., [42, 43]), which was firstly defined for S-MJLSs in [39], assuming that each sojourn-time is truncated by the given upper

\* Corresponding author (email: tangtany@gmail.com)

1) For convenience,  $\sigma$ -MSS will denote  $\sigma$ -error mean square stability or  $\sigma$ -error mean square stable in the following.

bound. In [40], the stability and stabilization were investigated for discrete-time S-MJLSs with general and exponentially periodic distributions of sojourn-time, respectively.

The semi-Markov kernel is usually calculated using S-TPMFs and the transition probability matrix of the embedded chain. Thus, in order to tackle the stability problem for discrete-time S-MJLSs with the semi-Markov kernel approach, almost all existing results assume that specific S-TPMFs are available. However, in practice, it is difficult to obtain accurate and comprehensive information for the S-TPMFs, because they are usually deduced from statistical characteristics according to the sampled data. In fact, adequate samples for random sojourn-time and mode transitions of semi-Markov chains are costly and time-consuming [44]; inadequate samples may result in an incorrect presumption for S-TPMFs, which leads to the designed controllers failing to stabilize the S-MJLS. However, the stability criteria established for discrete-time S-MJLSs with the semi-Markov kernel approach, are usually dependent on the given upper bound of sojourn-time, which significantly affects the practical use of derived results [44]. This leads to the fact that the results in [39,42,43] are not applicable when the specific information of S-TPMFs is unavailable.

Motivated by the aforementioned discussions, this study addresses the stability problem for discrete-time S-MJLSs with unavailable S-TPMFs information. The major contributions of this study are as follows:

(1) The stability problem is investigated for discrete-time S-MJLSs without using any S-TPMFs information, where only the transition probability matrix of embedded chains is used for semi-Markov chain. The results in [27, 39, 40, 43] depend heavily on the S-TPMFs information, which cannot be applied to the situation considered in this study.

(2) A novel analysis technique is developed to study the  $\sigma$ -MSS for discrete-time S-MJLSs under some widely used assumptions, and the established stability criteria remove the dependence of the upper bound for sojourn-time in [39, 40, 42, 43, 45].

The remainder of this paper is organized as follows. The model is introduced in Section 2, along with some necessary definitions. Section 3 is divided into two parts. The stability criteria are established for unforced discrete-time S-MJLSs in the first part, and in the second part, the existence conditions for stabilizing the controller are presented for closed-loop S-MJLSs. Finally, two numerical examples are provided to demonstrate the validity of the obtained theoretical results.

**Notations.** Set  $\mathbb{N}^+ = \{1, 2, \dots\}$ ,  $\mathbb{N} = \mathbb{N}^+ \cup \{0\}$ , and  $\Gamma = \{1, 2, \dots, \vartheta\}$ , where  $\vartheta$  is a finite positive integer. The  $d$ -dimensional Euclidean space is represented by  $\mathbb{R}^d$ .  $\mathbb{N}_{[n_1, n_2]}$  represents the set  $\{n \in \mathbb{N} | n_1 \leq n \leq n_2\}$ .  $(\Omega, \mathcal{F}, \mathbb{P})$  is a complete probability space, and  $\mathbb{E}$  is the mathematical expectation defined in this space.  $P'$  and  $P^{-1}$  represent the transposition and inverse of matrix  $P$ , respectively.  $\text{diag}\{\dots\}$  is defined as a block-diagonal matrix, and the asterisk  $*$  is employed as an ellipsis for terms introduced by symmetry.  $\lambda_{\min}(P)$  ( $\lambda_{\max}(P)$ ) describes the smallest (largest) eigenvalue of matrix  $P$ .  $\text{diag}_{(\vartheta)}\{P\}$  is defined as a  $\vartheta \times \vartheta$  block-diagonal matrix, whose diagonal entries are  $P$ . Notation  $P \succ 0$  ( $P \prec 0$ ) denotes the matrix  $P$  is positive (negative) definite and real symmetric. The symbols  $\underline{\text{Ge}}(a_1)$ ,  $\underline{\text{P}}(a_2)$ , and  $\underline{\text{b}}(a_3, a_4)$  denote the geometric distribution with parameter  $a_1$ , Poisson distribution with parameters  $a_2$ , and binomial distribution with parameters  $a_3$  and  $a_4$ , respectively. The mathematical expectation of  $x$  conditional on  $y$  is defined as  $\mathbb{E}x|y$ . The symbol  $I(\cdot)$  denotes the indicator function.

## 2 Preliminaries

Consider the following discrete-time S-MJLS:

$$x(n+1) = A_{r(n)}x(n) + B_{r(n)}u(n), \quad n \in \mathbb{N}, \quad (1)$$

where  $x(n) \in \mathbb{R}^{d_x}$  is the system state,  $\{r(n)\}_{n \in \mathbb{N}}$  is a semi-Markov chain taking values from the finite set  $\Gamma$ , and  $u(n) \in \mathbb{R}^{d_u}$  is the control input. Both  $A_i$  and  $B_i$  are defined as real matrices with appropriate dimensions for any  $i \in \Gamma$ . For simplicity, let initial mode  $r(0) = r_0$  and the initial state  $x(0) = x_0$ .

To introduce the definition of a semi-Markov chain, some necessary notations and definitions are provided first. Let  $\varrho_n$  and  $\zeta_n$ , respectively, be the time instant and the index of the subsystem mode at the  $n$ th jump of the stochastic process  $\{r(n)\}_{n \in \mathbb{N}}$ . The sojourn-time between the  $n$ th and  $(n+1)$ th jump is defined as  $\eta_n$ , where  $\eta_n = \varrho_{n+1} - \varrho_n$  for  $n \in \mathbb{N}$  and  $\varrho_0 = 0$ . The Markov renewal chain (MRC) is defined as follows.

**Definition 1** ([40]). If  $\{(\zeta_n, \varrho_n)\}_{n \in \mathbb{N}}$  is a discrete-time stochastic process, and for  $\forall k \in \mathbb{N}^+$  and  $i, j \in \Gamma$ ,

$$\begin{aligned} \mathbb{P}(\zeta_{n+1} = j, \eta_n = k | \zeta_n = i, \eta_{n-1}, \dots, \zeta_1, \eta_0, \zeta_0) &= \mathbb{P}(\zeta_{n+1} = j, \eta_n = k | \zeta_n = i) \\ &= \mathbb{P}(\zeta_1 = j, \eta_0 = k | \zeta_0 = i), \end{aligned} \tag{2}$$

then  $\{(\zeta_n, \varrho_n)\}_{n \in \mathbb{N}}$  is said to be a homogeneous MRC.

Stochastic process  $\{\zeta_n\}_{n \in \mathbb{N}}$  is a Markov chain, which is usually called the embedded chain of MRC  $\{(\zeta_n, \varrho_n)\}_{n \in \mathbb{N}}$ . Let  $\Theta = [\theta_{ij}]_{\vartheta \times \vartheta}$  be the transition probability matrix of  $\{\zeta_n\}_{n \in \mathbb{N}}$  with  $\theta_{ij} = \mathbb{P}(\zeta_{n+1} = j | \zeta_n = i)$  and  $\theta_{ii} = 0$  for  $\forall i, j$ . In this study, the probability distribution of the sojourn-time is assumed to have the following property.

**Assumption 1** ([15]). Let  $\{(\zeta_n, \varrho_n)\}_{n \in \mathbb{N}}$  be an MRC and satisfy that for  $\forall k \in \mathbb{N}^+$  and  $i, j \in \Gamma$ ,

$$\mathbb{P}(\eta_n = k | \zeta_{n+1} = j, \zeta_n = i) = \mathbb{P}(\eta_n = k | \zeta_n = i) = \mathbb{P}(\eta_0 = k | \zeta_0 = i),$$

that is, the probability distribution of sojourn-time depends only on the current mode, and is independent of the next one.

Under Assumption 1, the semi-Markov kernel can be defined as

$$\begin{aligned} \pi_{ij}(k) &\triangleq \mathbb{P}(\zeta_{n+1} = j, \eta_n = k | \zeta_n = i) = \mathbb{P}(\zeta_{n+1} = j | \zeta_n = i) \mathbb{P}(\eta_n = k | \zeta_{n+1} = j, \zeta_n = i) \\ &= \theta_{ij} \omega_i(k), \end{aligned} \tag{3}$$

where  $\omega_i(k) \triangleq \mathbb{P}(\eta_{n+1} = k | \zeta_n = i)$  and  $k \in \mathbb{N}^+$ .

**Remark 1.** In Assumption 1, the probability distribution of sojourn-time is assumed to be independent of the next system mode. It is worth noting that this assumption is widely adopted for considering the stability/stabilization of semi-Markov jump systems [22, 26, 33, 46, 47]. Obviously, under Assumption 1, if the distribution  $\omega_i(k)$  of sojourn-time is limited to the geometric distribution, then the defined semi-Markov chain becomes a discrete-time Markov chain.

With the help of the definition for MRC, we can define a semi-Markov chain as follows.

**Definition 2** ([48]). Let  $\{(\zeta_n, \varrho_n)\}_{n \in \mathbb{N}}$  be an MRC, the stochastic process  $\{r(n)\}_{n \in \mathbb{N}}$  is said to be a semi-Markov chain, if for  $\forall k \in \mathbb{N}$ ,  $r(k) = \zeta_{N(k)}$ , where  $N(k) \triangleq \max\{n \in \mathbb{N} | \varrho_n \leq k\}$  is a discrete-time counting process.

A stability concept for the unforced system (1) can be provided, that is, the control input  $u(n) \equiv 0$  in system (1).

**Definition 3** ([39]). Given the upper bound  $\mathcal{T}_i$  of sojourn-time for the  $i$ th mode, the unforced system (1) is said to be  $\sigma$ -MSS, if for any initial value  $x_0 \in \mathbb{R}$ , the initial mode  $r_0 \in \Gamma$ ,  $i \in \Gamma$ , and  $\iota \in \mathbb{N}^+$ ,

$$\lim_{n \rightarrow \infty} \mathbb{E}[\|x(n)\|^2]_{|x_0, r_0, \eta_\iota I_{(\zeta_\iota = i)} \leq \mathcal{T}_i} = 0,$$

where  $\sigma \triangleq \sum_{i \in \Gamma} |\ln F_i(\mathcal{T}_i)|$  denotes the approximation error component of the  $i$ th mode and  $F_i(\mathcal{T}_i) = P(\eta_\iota \leq \mathcal{T}_i | \zeta_\iota = i)$ .

**Remark 2.** In most existing literature on  $\sigma$ -MSS, the error  $\sigma$  should be calculated to characterize the approximate effect. In this study, it is unnecessary because the error  $\sigma$  can always be sufficiently small. As pointing out in Remark 3 of [39], if  $\mathcal{T}_i \rightarrow \infty$  for each mode  $i \in \Gamma$ , then  $\sigma \rightarrow 0$ . In this study, the stability criteria presented do not depend on the given upper bound  $\mathcal{T}_i$ , we can always choose a sufficiently large  $\mathcal{T}_i$  such that the corresponding error  $\sigma$  is sufficiently small.

**Remark 3.** Many important results have been established for discrete-time S-MJLSs in [39, 40, 42, 43, 45]. However, the stability criteria in [39, 40, 42, 43, 45] cannot be used for discrete-time S-MJLSs, when the S-TPMFs information is unavailable. However, as pointed out in Remark 3 of [44], sufficient conditions provided in [39, 40, 42, 43, 45] depend on the given upper bound of sojourn-time. In this study, by assuming that the probability distribution of sojourn-time depends only on the current mode,  $\sigma$ -MSS is presented for discrete-time S-MJLSs without using any S-TPMF information, and the stability criteria are not affected by the exactly truncated upper bounds of sojourn-time.

### 3 Main results

In this section, the criteria of  $\sigma$ -MSS are presented for the unforced system (1) first, and thereafter the existence conditions of state-feedback stabilizing controllers are provided such that the resulting closed-loop system (1) is  $\sigma$ -MSS, where the sojourn-time information is assumed to be unavailable.

### 3.1 MSS for unforced systems

In this subsection, stability criteria were established for the unforced S-MJLS (1). Before proceeding further, a lemma is presented as follows:

**Lemma 1.** For unforced S-MJLS (1), if there exists a set of positive definite matrices  $P_i$  such that for  $\forall i \in \Gamma$ ,

$$(H_1) \quad A'_i \mathcal{P}^i A_i - P_i \prec 0,$$

$$(H_2) \quad A'_i \mathcal{P}^i A_i - \mathcal{P}^i \prec 0,$$

where  $\mathcal{P}^i = \sum_{j \in \Gamma} \theta_{ij} P_j$ , then

$$\lim_{n \rightarrow \infty} \mathbb{E} \|x(\varrho_n)\|^2 = 0. \tag{4}$$

*Proof.* Let the Lyapunov function

$$V_i(x(n)) = x'(n) P_i x(n),$$

where  $i \in \Gamma$  and  $n \in \mathbb{N}$ . For  $i, j \in \Gamma$  and  $k_n \in \mathbb{N}$ , using (3), one has

$$\mathbb{P}(r(\varrho_{n+1}) = j, \eta_n = k_n | r(\varrho_n) = i) = \theta_{ij} \mathbb{P}(\eta_n = k_n | r(\varrho_n) = i). \tag{5}$$

From (5), it is not difficult to see that

$$\begin{aligned} & \mathbb{E}[V_{r(\varrho_{n+1})}(x(\varrho_{n+1})) | x(\varrho_n), r(\varrho_n) = i] \\ &= \sum_{k_n=1}^{+\infty} \sum_{j \in \Gamma} x'(\varrho_{n+1}) P_j x(\varrho_{n+1}) I_{(\eta_n=k_n)} \mathbb{P}(r(\varrho_{n+1}) = j, \eta_n = k_n | r(\varrho_n) = i) \\ &= \sum_{k_n=1}^{+\infty} \sum_{j \in \Gamma} x'(\varrho_{n+1}) P_j x(\varrho_{n+1}) I_{(\eta_n=k_n)} \mathbb{P}(\eta_n = k_n | r(\varrho_n) = i) \theta_{ij}. \end{aligned}$$

It follows that

$$\mathbb{E}[V_{r(\varrho_{n+1})}(x(\varrho_{n+1})) | x(\varrho_n), r(\varrho_n) = i] = \sum_{k_n=1}^{+\infty} \mathbb{P}(\eta_n = k_n | r(\varrho_n) = i) I_{(\eta_n=k_n)} x'(\varrho_{n+1}) \mathcal{P}^i x(\varrho_{n+1}),$$

where  $\mathcal{P}^i = \sum_{j \in \Gamma} \theta_{ij} P_j$ . Noticing the form of unforced S-MJLS (1), we have

$$\mathbb{E}[V_{r(\varrho_{n+1})}(x(\varrho_{n+1})) | x(\varrho_n), r(\varrho_n) = i] = \sum_{k_n=1}^{+\infty} P(\eta_n = k_n | r(\varrho_n) = i) x'(\varrho_n) (A'_i)^{k_n} \mathcal{P}^i A_i^{k_n} x(\varrho_n). \tag{6}$$

In view of (H<sub>2</sub>), we derive that for any integral  $\iota \geq 2$ ,

$$(A'_i)^\iota \mathcal{P}^i A_i^\iota - (A'_i)^{\iota-1} \mathcal{P}^i A_i^{\iota-1} \prec 0,$$

which yields that

$$\begin{aligned} & (A'_i)^{k_n} \mathcal{P}^i A_i^{k_n} - A'_i \mathcal{P}^i A_i \\ &= \left( (A'_i)^{k_n} \mathcal{P}^i A_i^{k_n} - (A'_i)^{k_n-1} \mathcal{P}^i A_i^{k_n-1} \right) + \left( (A'_i)^{k_n-2} \mathcal{P}^i A_i^{k_n-2} - (A'_i)^{k_n-3} \mathcal{P}^i A_i^{k_n-3} \right) \\ & \quad + \cdots + \left( (A'_i)^2 \mathcal{P}^i A_i^2 - (A'_i) \mathcal{P}^i A_i \right) \\ &= \sum_{\iota=2}^{k_n} \left[ (A'_i)^\iota \mathcal{P}^i A_i^\iota - (A'_i)^{\iota-1} \mathcal{P}^i A_i^{\iota-1} \right] \prec 0. \end{aligned} \tag{7}$$

Substituting (7) into (6) yields that

$$\mathbb{E}[V_{r(\varrho_{n+1})}(x(\varrho_{n+1})) | x(\varrho_n), r(\varrho_n) = i] \leq \sum_{k_n=1}^{+\infty} \mathbb{P}(\eta_n = k_n | r(\varrho_n) = i) x'(\varrho_n) A'_i \mathcal{P}^i A_i x(\varrho_n)$$

$$=x'(\varrho_n)A_i'P^iA_ix(\varrho_n), \tag{8}$$

where  $\sum_{k_n=1}^{+\infty} \mathbb{P}(\eta_n = k_n|r(\varrho_n) = i) = 1$ . It follows from (H<sub>1</sub>) that  $\beta \triangleq \min_{i \in \Gamma} \{\lambda_{\min}(-A_i'P^iA_i + P_i)\} > 0$ . Noticing (H<sub>1</sub>) and (8), for any  $r(\varrho_n) = i \in \Gamma$ , we have

$$\begin{aligned} \mathbb{E}[V_{r(\varrho_{n+1})}(x(\varrho_{n+1}))|x(\varrho_n), r(\varrho_n) = i] - V_i(x(\varrho_n)) &\leq x'(\varrho_n)A_i'P^iA_ix(\varrho_n) - x'(\varrho_n)P_ix(\varrho_n) \\ &= x'(\varrho_n)(A_i'P^iA_i - P_i)x(\varrho_n) \\ &\leq -\beta\|x(\varrho_n)\|^2. \end{aligned}$$

Consequently, for any  $r(\varrho_n) \in \Gamma$  and  $n \in \mathbb{N}$ ,

$$\beta\|x(\varrho_n)\|^2 \leq [V_{r(\varrho_n)}(x(\varrho_n)) - \mathbb{E}[V_{r(\varrho_{n+1})}(x(\varrho_{n+1}))|x(\varrho_n), r(\varrho_n)]] . \tag{9}$$

By taking the expectation on the both sides of (9), one has

$$\mathbb{E}\|x(\varrho_n)\|^2 \leq \frac{1}{\beta} [\mathbb{E}V_{r(\varrho_n)}(x(\varrho_n)) - \mathbb{E}V_{r(\varrho_{n+1})}(x(\varrho_{n+1}))] . \tag{10}$$

It follows from (10) that for  $\forall n \in \mathbb{N}$ ,

$$\sum_{\iota=0}^n \mathbb{E}\|x(\varrho_\iota)\|^2 \leq \frac{1}{\beta} [\mathbb{E}V_{r(0)}(x(0)) - \mathbb{E}V_{r(\varrho_{n+1})}(x(\varrho_{n+1}))] \leq \frac{1}{\beta} \mathbb{E}V_{r(0)}(x(0)) < +\infty, \tag{11}$$

which implies that

$$\lim_{n \rightarrow \infty} \mathbb{E}\|x(\varrho_n)\|^2 = 0. \tag{12}$$

Based on Lemma 1, sufficient conditions will be given to ensure that the unforced S-MJLS (1) is  $\sigma$ -MSS.

**Theorem 1.** If there exist positive constants  $\kappa_i, \mathcal{T}_i$ , and a set of positive definite matrices  $P_i$  such that for  $\forall i \in \Gamma$ , (H<sub>1</sub>) and (H<sub>2</sub>) hold, and

$$(H_3) \ A_i'P_iA_i - \kappa_iP_i \prec 0,$$

then unforced S-MJLS (1) is  $\sigma$ -MSS.

*Proof.* For any  $i \in \Gamma$  and  $n \in \mathbb{N}$ , let the Lyapunov function

$$V_i(x(n)) = x'(n)P_ix(n).$$

If  $r(\varrho_n) = i \in \Gamma$ , and  $\mathcal{T}_i$  is the given upper bound of the sojourn-time for the  $i$ th mode, then for  $\forall \bar{n} \in \mathbb{N}_{[\varrho_n+1, \varrho_{n+1}]}$ , we have  $\bar{n} - \varrho_n \leq \mathcal{T}_i$ . Using (H<sub>3</sub>), we have

$$\begin{aligned} V_{r(\varrho_n)}(x(\bar{n})) &= x'(\bar{n})P_ix(\bar{n}) = x'(\varrho_n)(A_i')^{\bar{n}-\varrho_n}P_iA_i^{\bar{n}-\varrho_n}x(\varrho_n) \leq \kappa_i^{\bar{n}-\varrho_n}V_{r(\varrho_n)}(x(\varrho_n)) \\ &\leq \lambda_{\max}(P_i)(\bar{\kappa}_i)^{\mathcal{T}_i}\|x(\varrho_n)\|^2, \end{aligned} \tag{13}$$

where  $\bar{\kappa}_i = \max\{\kappa_i, 1\}$  and  $r(\varrho_n) = i$ . It follows that for  $\bar{n} \in \mathbb{N}_{[\varrho_n+1, \varrho_{n+1}]}$ ,

$$V_{r(\varrho_n)}(x(\bar{n})) \leq c_1\|x(\varrho_n)\|^2,$$

where  $c_1 = \max_{i \in \Gamma} \{\lambda_{\max}(P_i)(\bar{\kappa}_i)^{\mathcal{T}_i}\}$ . Consequently,

$$\mathbb{E}V_{r(\varrho_n)}(x(\bar{n})) \leq c_1\mathbb{E}\|x(\varrho_n)\|^2. \tag{14}$$

In addition, it can be verified that

$$c_2\mathbb{E}\|x(\bar{n})\|^2 \leq \mathbb{E}V_{r(\varrho_n)}(x(\bar{n})),$$

where  $c_2 = \min_{i \in \Gamma} \{\lambda_{\min}(P_i)\}$ . Thus, for  $\forall \bar{n} \in \mathbb{N}$ , there is a corresponding  $n$  such that  $\bar{n} \in \mathbb{N}_{[\varrho_n+1, \varrho_{n+1}]}$  and

$$\mathbb{E}\|x(\bar{n})\|^2 \leq \frac{c_1}{c_2}\mathbb{E}\|x(\varrho_n)\|^2. \tag{15}$$

Combining Lemma 1 with (15) yields that

$$\lim_{\bar{n} \rightarrow \infty} \mathbb{E}\|x(\bar{n})\|^2|_{x_0, r_0, \eta_n I_{(\zeta_n=i)} \leq \mathcal{T}_i} = 0.$$

It follows that unforced S-MJLS (1) is  $\sigma$ -MSS.

**Remark 4.** Conditions (H<sub>1</sub>)–(H<sub>3</sub>) in Theorem 1 are commonly used to study the stability of discrete-time Markov/semi-Markov jump linear systems. Condition (H<sub>1</sub>) is a sufficient and necessary condition to obtain the mean square stability (MSS) for unforced system (1) by assuming  $\{r(n)\}_{n \in \mathbb{N}}$  is a Markov chain [5]. Condition (H<sub>2</sub>) is similar to some assumptions, which are necessary to handle the stability for discrete-time S-MJLS with a semi-Markov kernel approach, for example, assumption (16) in [40] and assumption (31) in [43]. Condition (H<sub>3</sub>) establishes the relationship between  $V_{r(\varrho_n)}(x(\bar{n}))$  and  $V_{r(\varrho_n)}(x(\varrho_n))$  for any  $\bar{n} \in \mathbb{N}_{[\varrho_n, \varrho_{n+1}]}$ , which was also utilized in [39, 40, 43].

Our next result provides a different set of conditions to ensure that the unforced S-MJLS (1) is  $\sigma$ -MSS.

**Theorem 2.** If there exist constants  $\kappa_i > 0, \mathcal{T}_i > 0, \mu_i > -1$ , and a set of positive definite matrices  $P_i$  such that for  $\forall i \in \Gamma$ , (H<sub>2</sub>) and (H<sub>3</sub>) in Theorem 1 hold, and

$$(H_4) \quad A_i' \mathcal{P}^i A_i - (1 + \mu_i) P_i \prec 0,$$

$$(H_5) \quad \max_{i \in \Gamma} \sum_{j \in \Gamma} (1 + \mu_i) \theta_{ij} < 1,$$

where  $\mathcal{P}^i = \sum_{j \in \Gamma} \theta_{ij} P_j$ , then unforced S-MJLS (1) is  $\sigma$ -MSS.

*Proof.* For  $\forall i \in \Gamma$  and  $n \in \mathbb{N}$ , let the Lyapunov function

$$V_i(x(n)) = x'(n) P_i x(n).$$

Set  $\mathcal{F}_n = \sigma(\{x(\varrho_k)\}_{k \in \mathbb{N}_{[0, n]}}, \{r(\varrho_k)\}_{k \in \mathbb{N}_{[0, n]}})$  with  $n \in \mathbb{N}$ , which is a  $\sigma$ -algebra generated by  $\{x(\varrho_k)\}_{k \in \mathbb{N}_{[0, n]}}$  and  $\{r(\varrho_k)\}_{k \in \mathbb{N}_{[0, n]}}$ . It can be checked that

$$\mathbb{E}[V_{r(\varrho_{n+1})}(x(\varrho_{n+1})) | \mathcal{F}_n] = \mathbb{E}[V_{r(\varrho_{n+1})}(x(\varrho_{n+1})) | x(\varrho_n), r(\varrho_n)]. \tag{16}$$

If  $r(\varrho_n) = i$ , then using (8) in the proof of Lemma 1 and (H<sub>4</sub>), we have

$$\mathbb{E}[V_{r(\varrho_{n+1})}(x(\varrho_{n+1})) | x(\varrho_n), r(\varrho_n) = i] \leq x'(\varrho_n) A_i' \mathcal{P}^i A_i x(\varrho_n) \leq (1 + \mu_i) V_i(x(\varrho_n)), \quad n \in \mathbb{N}.$$

It follows that for  $\forall n \in \mathbb{N}$ ,

$$\mathbb{E}[V_{r(\varrho_{n+1})}(x(\varrho_{n+1})) | x(\varrho_n), r(\varrho_n)] \leq (1 + \mu_{r(\varrho_n)}) V_{r(\varrho_n)}(x(\varrho_n)). \tag{17}$$

In view of (16) and (17), for  $\forall n \in \mathbb{N}$ , one has

$$\mathbb{E}[V_{r(\varrho_{n+1})}(x(\varrho_{n+1})) | \mathcal{F}_n] \leq (1 + \mu_{r(\varrho_n)}) V_{r(\varrho_n)}(x(\varrho_n)). \tag{18}$$

By utilizing the definition of  $\mathcal{F}_n$  and (18), we derive that

$$\begin{aligned} \mathbb{E}[V_{r(\varrho_{n+1})}(x(\varrho_{n+1})) | \mathcal{F}_{n-1}] &= \mathbb{E}[\mathbb{E}[V_{r(\varrho_{n+1})}(x(\varrho_{n+1})) | \mathcal{F}_n] | \mathcal{F}_{n-1}] \\ &\leq (1 + \mu_{r(\varrho_n)}) \mathbb{E}[V_{r(\varrho_n)}(x(\varrho_n)) | \mathcal{F}_{n-1}] \\ &\leq \prod_{k=n-1}^n (1 + \mu_{r(\varrho_k)}) V_{r(\varrho_{n-1})}(x(\varrho_{n-1})). \end{aligned}$$

Repeating this procedure yields that

$$\mathbb{E}[V_{r(\varrho_{n+1})}(x(\varrho_{n+1})) | \mathcal{F}_0] \leq V_{r_0}(x_0) \prod_{k=0}^n (1 + \mu_{r(\varrho_k)}).$$

Thus, we obtain that

$$\mathbb{E}[V_{r(\varrho_{n+1})}(x(\varrho_{n+1}))] \leq (1 + \mu_{r_0}) V_{r_0}(x_0) \mathbb{E} \left[ \prod_{k=1}^n (1 + \mu_{r(\varrho_k)}) \right]. \tag{19}$$

It can be verified that

$$\begin{aligned} \mathbb{E} \left[ \prod_{k=1}^n (1 + \mu_{r(\varrho_k)}) \right] &= \sum_{i_1, \dots, i_n} \prod_{k=1}^n (1 + \mu_{i_k}) \mathbb{P}(r(\varrho_1) = i_1, \dots, r(\varrho_n) = i_n) \\ &= \sum_{i_1, \dots, i_n} \prod_{k=1}^n (1 + \mu_{i_k}) \theta_{i_{k-1} i_k} = \prod_{k=1}^n \sum_{i_k} (1 + \mu_{i_k}) \theta_{i_{k-1} i_k} \end{aligned}$$

$$\leq \varpi^n, \tag{20}$$

where  $\varpi = \max_{i \in \Gamma} \sum_{j \in \Gamma} (1 + \mu_i) \theta_{ij}$  and  $i_0 = r_0$ . Combining (19) and (20) with (H<sub>5</sub>) yields that

$$\lim_{n \rightarrow \infty} \mathbb{E}[V_{r(\varrho_n)}(x(\varrho_n))] = 0. \tag{21}$$

Using (H<sub>3</sub>), we can obtain (15) as in the proof of Theorem 1. Thereafter, for the given  $\mathcal{T}_i$ , in view of (21) and (15), one has

$$\lim_{n \rightarrow \infty} \mathbb{E}\|x(n)\|^2 |_{x_0, r_0, \eta_n I_{(\zeta_n=i)} \leq \mathcal{T}_i} = 0.$$

It follows that unforced S-MJLS (1) is  $\sigma$ -MSS.

**Remark 5.** It can be verified that (H<sub>4</sub>) in Theorem 2 is an extension of (H<sub>1</sub>) in Theorem 1, where (H<sub>4</sub>) becomes (H<sub>1</sub>) by letting  $\mu_i \equiv 0$ . The  $\sigma$ -MSS was presented in [43] for S-MJLS using the semi-Markov kernel approach. Compared with Theorem 3.3 in [43], a novel analysis method is introduced in Theorem 2 to study the  $\sigma$ -MSS for S-MJLSs, and there are no specific requirements for the information of S-TPMFs and the given upper bound of sojourn-time.

**Remark 6.** It is known that (H<sub>1</sub>) is a sufficient and necessary condition to obtain the MSS for discrete-time Markov jump linear systems (see [5]), where the geometric distribution information is not assumed. In this paper,  $\sigma$ -MSS is investigated for S-MJLSs, where the sojourn-time information is also not required. Compared with the classical result in [5], this paper has the following two contributions:

(1) Markov chain is extended to semi-Markov chain. In addition, (H<sub>1</sub>) is further generalized to (H<sub>4</sub>) in Theorem 2, where  $\sigma$ -MSS is studied for S-MJLSs under the general condition (H<sub>4</sub>);

(2) A novel analysis technique is developed in this study, which combines the Lyapunov function analysis method with analysis technique of Markov chains.

**Remark 7.** Numerous important results have been reported for the stability analysis and control problem for S-MJLS in [27, 28, 39–43] by using the semi-Markov kernel approach, and the stability criteria in [39, 40, 42, 43, 45] depend on the given upper bound of sojourn-time. However, the S-TPMFs information is difficult to obtain in practice. This is because S-TPMFs are usually deduced from the statistical characteristics according to the sampled-data, while adequate samples are often costly and time consuming. In addition, the stability results depend on the upper bound of sojourn-time, which seriously affects their application [44]. Thus, in this study,  $\sigma$ -MSS was considered without using the S-TPMFs information. The main contributions of Theorems 1 and 2 lie in the following two parts:

(1) Under some widely used assumptions, the stability criteria are provided without using any S-TPMFs information, which means that the results in Theorems 1 and 2, are independent of the specific form for S-TPMFs.

(2) The exactly truncated upper bound of the sojourn-time is not embodied in the stability criteria.

### 3.2 Controller design for closed-loop systems

Based on Theorems 1 and 2, the sufficient conditions of stabilizing controller are presented for S-MJLS (1) in this subsection. In this study, the mode-dependent controller is assumed to have the following form:

$$u(n) = K_{r(n)}x(n), \tag{22}$$

where  $r(n) = i \in \Gamma$ ,  $n \in \mathbb{N}$  and  $K_i$  is the control gain to be determined. Therefore, we obtain the following closed-loop S-MJLS:

$$x(n+1) = \bar{A}_{r(n)}x(n), \tag{23}$$

where  $\bar{A}_{r(n)} = A_{r(n)} + B_{r(n)}K_{r(n)}$  and  $n \in \mathbb{N}$ . In the following, effective state-feedback gains are designed for the closed-loop system (23) with the help of Theorems 1 and 2, respectively.

**Theorem 3.** Assuming that there exist constants  $\kappa_i \in \mathbb{R}^+$  and matrices  $L_i, Q_i, U_i$  for  $i \in \Gamma$  such that the following inequalities hold:

$$\begin{bmatrix} L_i - Q'_i - Q_i & A_i Q_i + B_i U_i \\ * & -\kappa_i L_i \end{bmatrix} \prec 0, \tag{24}$$



$$\begin{bmatrix} \Upsilon_i (\mathcal{A}_i \mathcal{Q}_i + \mathcal{B}_i \mathcal{U}_i) R_i \\ * & L_i \end{bmatrix} \prec 0, \quad (25)$$

$$\begin{bmatrix} \Upsilon_i (\mathcal{A}_i \mathcal{Q}_i + \mathcal{B}_i \mathcal{U}_i) R_i \\ * & \bar{L}_i \end{bmatrix} \prec 0, \quad (26)$$

where

$$\begin{aligned} \Upsilon_i &= \mathcal{L} - \mathcal{Q}_i - \mathcal{Q}'_i, \quad \mathcal{Q}_i = \text{diag}_{(\vartheta)}\{Q_i\}, \quad \mathcal{L} = \text{diag}\{L_1, L_2, \dots, L_\vartheta\}, \\ \bar{L}_i &= \sum_{j \in \Gamma} \theta_{ij} L_j, \quad R_i = (\sqrt{\theta_{i1}} I, \sqrt{\theta_{i2}} I, \dots, \sqrt{\theta_{i\vartheta}} I)', \quad \mathcal{A}_i = \text{diag}_{(\vartheta)}\{A_i\}, \end{aligned}$$

$\mathcal{B}_i = \text{diag}_{(\vartheta)}\{B_i\}$ , and  $\mathcal{U}_i = \text{diag}_{(\vartheta)}\{U_i\}$ . Then, the closed-loop system (23) is  $\sigma$ -MSS with state-feedback gain  $K_i = U_i Q_i^{-1}$  for  $\forall i \in \Gamma$ .

*Proof.* In this proof, it will be shown that conditions (24)–(26) can imply (H<sub>1</sub>)–(H<sub>3</sub>) in Theorem 1 for the closed-loop system (23). If we set

$$Y_i = Q_i^{-1}, \quad P_i = (Q'_i)^{-1} L_i Q_i^{-1}, \quad K_i = U_i Q_i^{-1},$$

then it can be deduced from (24) that for  $\forall i \in \Gamma$ ,

$$\begin{bmatrix} (Y'_i)^{-1} P_i Y_i^{-1} - (Y'_i)^{-1} - Y_i^{-1} & A_i Y_i^{-1} + B_i K_i Y_i^{-1} \\ * & -\kappa_i (Y'_i)^{-1} P_i Y_i^{-1} \end{bmatrix} \prec 0. \quad (27)$$

Taking a congruence transformation to (27) with  $\text{diag}\{Y_i, Y_i\}$ , it follows that

$$\begin{bmatrix} P_i - Y_i - Y'_i & Y'_i (A_i + B_i K_i) \\ * & -\kappa_i P_i \end{bmatrix} \prec 0. \quad (28)$$

In view of  $(P_i - Y_i)' P_i^{-1} (P_i - Y_i) \succ 0$ , we have  $-Y'_i P_i^{-1} Y_i \prec P_i - Y_i - Y'_i$ . Thereafter, it can be derived from (28) that

$$\begin{bmatrix} -Y'_i P_i^{-1} Y_i & Y'_i (A_i + B_i K_i) \\ * & -\kappa_i P_i \end{bmatrix} \prec 0. \quad (29)$$

By taking a congruence transformation to (29) with  $\text{diag}\{Y_i^{-1} P_i, I\}$ , we obtain

$$\begin{bmatrix} -P_i & P_i (A_i + B_i K_i) \\ * & -\kappa_i P_i \end{bmatrix} \prec 0. \quad (30)$$

Combining (30) with Schur complement follows (H<sub>3</sub>).

Let

$$\mathcal{Y} = \text{diag}\{Y_1, Y_2, \dots, Y_\vartheta\}, \quad \Xi = \text{diag}\{P_1, P_2, \dots, P_\vartheta\}, \quad \mathcal{K}_i = U_i Q_i^{-1}.$$

It follows that  $\mathcal{Y} = \mathcal{Q}^{-1}$  and  $\Xi = (\mathcal{Q}')^{-1} \mathcal{L} \mathcal{Q}^{-1}$ . Thereafter, it can be verified from (25) that

$$\begin{bmatrix} \Psi (\mathcal{A}_i + \mathcal{B}_i \mathcal{K}_i) \mathcal{Q}_i R_i \\ * & -L_i \end{bmatrix} \prec 0, \quad (31)$$

where  $\Psi = (\mathcal{Y}')^{-1} \Xi \mathcal{Y}^{-1} - (\mathcal{Y}')^{-1} - \mathcal{Y}^{-1}$ . Applying the congruence transformation to (31) with  $\text{diag}\{\mathcal{Y}, Y_i\}$ , we obtain

$$\begin{bmatrix} \Xi - \mathcal{Y} - \mathcal{Y}' & \mathcal{Y}' (\mathcal{A}_i + \mathcal{B}_i \mathcal{K}_i) R_i \\ * & -Y'_i L_i Y_i \end{bmatrix} \prec 0.$$



Since  $(\Xi - \mathcal{Y})'\Xi^{-1}(\Xi - \mathcal{Y}) \succ 0$ , it follows that  $-\mathcal{Y}\Xi^{-1}\Xi - \mathcal{Y} \prec \Xi - \mathcal{Y} - \mathcal{Y}'$ . Thereafter,

$$\begin{bmatrix} -\mathcal{Y}\Xi^{-1}\mathcal{Y} & \mathcal{Y}'(\mathcal{A}_i + \mathcal{B}_i\mathcal{K}_i)R_i \\ * & -Y_i' L_i Y_i \end{bmatrix} \prec 0. \quad (32)$$

Utilizing the congruence transformation to (32) with  $\text{diag}\{\mathcal{Y}^{-1}\Xi, I\}$ , we obtain

$$\begin{bmatrix} -\Xi & \Xi(\mathcal{A}_i + \mathcal{B}_i\mathcal{K}_i)R_i \\ * & -(Q_i')^{-1}L_i Q_i \end{bmatrix} \prec 0. \quad (33)$$

Noticing that Schur complement and (33), we obtain

$$R_i'(A_i + B_i K_i)'\Xi(A_i + B_i K_i)R_i - P_i \prec 0, \quad (34)$$

where  $P_i = (Q_i')^{-1}L_i Q_i$ . It can be verified that Eq. (34) is equivalent to (H<sub>1</sub>). Similarly, it can be verified that Eq. (26) yields (H<sub>2</sub>). Thus, in view of Theorem 1, the closed-loop system (23) is  $\sigma$ -MSS with a designed state-feedback gain  $K_i = U_i Q_i^{-1}$  for  $\forall i \in \Gamma$ .

The last result in this study deals with the stabilization of the closed system (23) by utilizing Theorem 2.

**Theorem 4.** Assuming that (H<sub>5</sub>) in Theorem 2 holds, and there exist constants  $\kappa_i \in \mathbb{R}^+$ ,  $\mu_i > -1$ , and matrices  $L_i, Q_i, U_i$  for  $i \in \Gamma$  such that the following inequalities hold:

$$\begin{bmatrix} L_i - Q_i' - Q_i & A_i Q_i + B_i U_i \\ * & -\kappa_i L_i \end{bmatrix} \prec 0, \quad (35)$$

$$\begin{bmatrix} \Upsilon_i & (A_i Q_i + B_i U_i)R_i \\ * & (1 + \mu_i)L_i \end{bmatrix} \prec 0, \quad (36)$$

$$\begin{bmatrix} \Upsilon_i & (A_i Q_i + B_i U_i)R_i \\ * & \bar{L}_i \end{bmatrix} \prec 0, \quad (37)$$

where

$$\begin{aligned} \Upsilon_i &= \mathcal{L} - \mathcal{Q}_i - \mathcal{Q}_i', \quad \mathcal{Q}_i = \text{diag}_{(\vartheta)}\{Q_i\}, \quad \mathcal{L} = \text{diag}\{L_1, L_2, \dots, L_\vartheta\}, \\ \bar{L}_i &= \sum_{j \in \Gamma} \theta_{ij} L_j, \quad R_i = (\sqrt{\theta_{i1}}I, \sqrt{\theta_{i2}}I, \dots, \sqrt{\theta_{i\vartheta}}I)', \quad \mathcal{A}_i = \text{diag}_{(\vartheta)}\{A_i\}, \end{aligned}$$

$\mathcal{B}_i = \text{diag}_{(\vartheta)}\{B_i\}$ , and  $\mathcal{U}_i = \text{diag}_{(\vartheta)}\{U_i\}$ . Then, the closed-loop system (23) is  $\sigma$ -MSS with state-feedback gain  $K_i = U_i Q_i^{-1}$  for  $\forall i \in \Gamma$ .

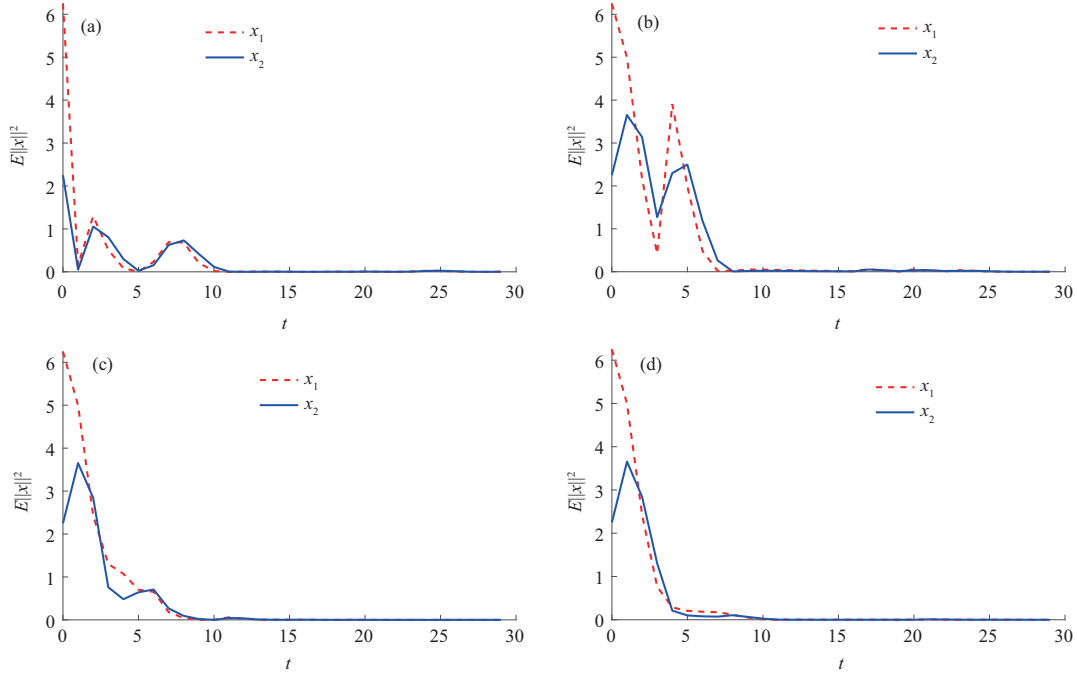
*Proof.* This analysis can be carried out along the same line as in the proof of Theorem 3.

## 4 Examples

Two numerical examples are presented in this section to validate the effectiveness of the obtained theoretical results.

**Example 1.** Assume that S-MJLS (1) has three subsystems, that is,  $\Gamma = \{1, 2, 3\}$ . The matrices in system (1) are defined as in [40] with

$$\begin{aligned} A_1 &= \begin{bmatrix} 1.55 & -0.85 \\ 0.50 & 0.15 \end{bmatrix}, \quad A_2 = \begin{bmatrix} 5.41 & -4.73 \\ 0.1 & 0.1 \end{bmatrix}, \quad A_3 = \begin{bmatrix} 0.27 & 0.50 \\ -1.31 & 2.03 \end{bmatrix}, \\ B_1 &= \begin{bmatrix} -0.1 \\ 0.1 \end{bmatrix}, \quad B_2 = \begin{bmatrix} 0.1 \\ 1.1 \end{bmatrix}, \quad B_3 = \begin{bmatrix} 0.1 \\ -0.1 \end{bmatrix}. \end{aligned}$$



**Figure 1** (Color online) (a) Average of 100 simulation results of state trajectories  $x(t)$  for system (1) in Example 1, where the S-TPMF satisfying Case 1 and the controller is designed using Theorem 3; (b) state trajectories  $x(t)$  for system (1) in Example 1, where the S-TPMF satisfying Case 2 and the controller is designed using Theorem 3; (c) state trajectories  $x(t)$  for system (1) in Example 1, where the S-TPMF satisfying Case 1 and the controller is designed using Theorem 4; (d) state trajectories  $x(t)$  for system (1) in Example 1, where the S-TPMF satisfying Case 2 and the controller is designed using Theorem 4.

The switching among these three modes is governed by a semi-Markov chain  $\{r(n)\}_{n \in \mathbb{N}}$ . Assume that this semi-Markov chain has a probability transition matrix

$$\Theta = \begin{bmatrix} 0 & 0.2262 & 0.7738 \\ 0.0912 & 0 & 0.9088 \\ 0.2463 & 0.7537 & 0 \end{bmatrix},$$

and S-TPMFs were unavailable. By solving the linear matrix inequality in Theorem 3, we can obtain the desired controller gains  $K_1 = [2.1541, -2.3942]$ ,  $K_2 = [4.5303, -3.9728]$ , and  $K_3 = [-3.5767, 2.2590]$ .

To demonstrate the validity of the results obtained in this study through simulation diagrams. Two different S-TPMFs cases are listed as follows:

Case 1.  $F_1 \sim \underline{\text{Ge}}(0.4)$ ,  $F_2 \sim \underline{\text{P}}(2)$ , and  $F_3 \sim \underline{\text{b}}(3, 0.2)$ .

Case 2.  $F_1 \sim \underline{\text{P}}(3)$ ,  $F_2 \sim \underline{\text{Ge}}(0.4)$ , and  $F_3 \sim \underline{\text{b}}(5, 0.4)$ .

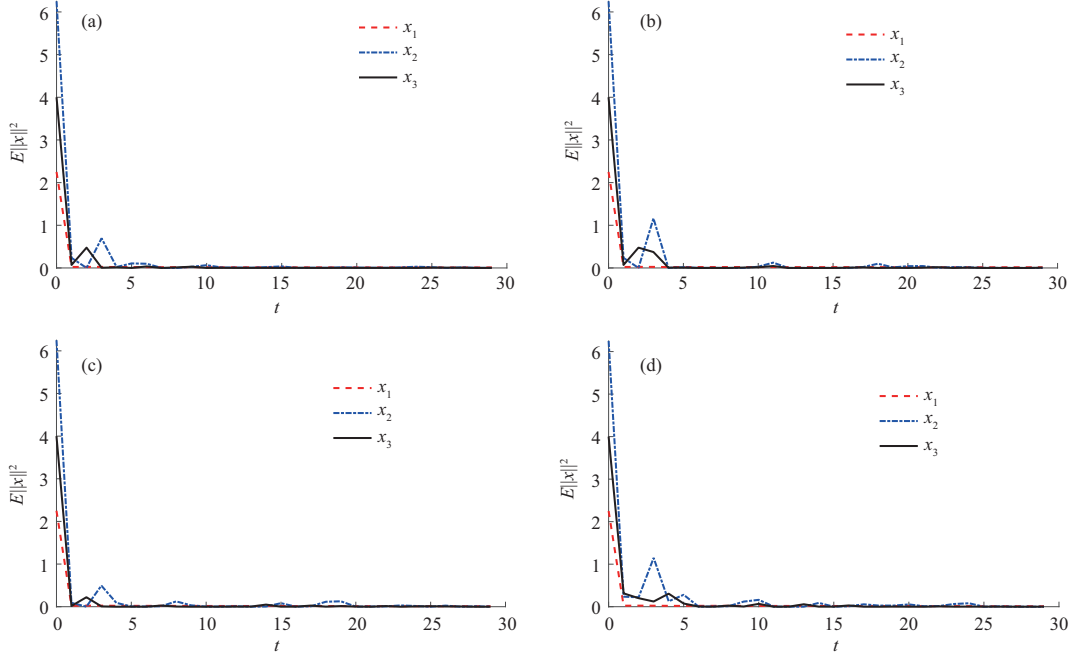
Thus, Assumption 1 is true for both Cases 1 and 2.

It is indicated in Figures 1(a) and (b) that S-MJLS (1) can be stabilized with the controller designed in Theorem 3, where the S-TPMFs are assumed to satisfy Cases 1 and 2, respectively. Thus, we find that the stability results in this study are independent of the specific form for S-TPMF.

Additionally, let  $\mu_1 = -0.9$ ,  $\mu_2 = -0.2$ , and  $\mu_3 = 0$  in Theorem 4. Using Theorem 4, we obtain the desired controller gains  $K_1 = [2.2230, -3.0042]$ ,  $K_2 = [4.5303, -3.9728]$ , and  $K_3 = [-3.3970, 2.4982]$ . Figures 1(c) and (d) show that S-MJLS (1) can be stabilized with the controller designed in Theorem 4, where the S-TPMFs are assumed to satisfy Cases 1 and 2, respectively.

In Example 1 of [39], S-MJLS (1) with a given S-TPMF for sojourn-time is presented as the  $\sigma$ -MSS. The results in [39] are dependent on the sojourn-time S-TPMFs, which are rarely used to design a controller for S-MJLS (1) without the specific S-TPMFs information. In this study, the controller is designed for S-MJLS (1) in [39], where only the probability transition matrix is assumed for semi-Markov chains. Figure 1 shows that the designed controller is effective for different S-TPMFs.

**Example 2.** In this example, a discrete-time electronic throttle control system is investigated as in [45], which can be expressed in the form of system (1). Let  $\chi(n)$ ,  $\psi(n)$ , and  $\omega(n)$  denote the angular position of the valve, angular velocity of the valve, and electrical current consumed by the internal motor of the



**Figure 2** (Color online) (a) Average of 100 simulation results of state trajectories  $x(t)$  for system (1) in Example 2, where the S-TPMF satisfying Case 1 and the controller is designed using Theorem 3; (b) state trajectories  $x(t)$  for system (1) in Example 2, where the S-TPMF satisfying Case 2 and the controller is designed using Theorem 3; (c) state trajectories  $x(t)$  for system (1) in Example 2, where the S-TPMF satisfying Case 1 and the controller is designed using Theorem 4; (d) state trajectories  $x(t)$  for system (1) in Example 2, where the S-TPMF satisfying Case 2 and the controller is designed using Theorem 4.

throttle [2], respectively. Thereafter, the system state vector is defined as  $x(n) = [\chi(n), \psi(n), \omega(n)]'$ . Assuming that the power amplifier has three operation modes: normal, soft failure, and hard failure, which correspond to  $r(n) = 1, 2$ , and  $3$ , respectively. For these three operation modes, the system parameters were defined as follows:

$$A_1 = \begin{bmatrix} 1 & 0.0109 & 0 \\ -0.1165 & 0.8072 & -1.5061 \\ 0 & -0.2285 & 0.7967 \end{bmatrix}, \quad A_2 = \begin{bmatrix} 1 & 0.0152 & 0 \\ -0.0298 & 0.2737 & -1.1921 \\ 0 & -0.3584 & 0.3835 \end{bmatrix},$$

$$A_3 = \begin{bmatrix} 1 & 0.0111 & 0 \\ -0.0229 & 0.7779 & -0.1899 \\ 0 & -0.6315 & 0.4178 \end{bmatrix}, \quad B_1 = \begin{bmatrix} 0 \\ 0 \\ -0.0948 \end{bmatrix}, \quad B_2 = \begin{bmatrix} 0 \\ 0 \\ -0.0735 \end{bmatrix}, \quad B_3 = \begin{bmatrix} 0 \\ 0 \\ -0.2441 \end{bmatrix}.$$

Assuming that the switching among the three modes satisfies a semi-Markov chain  $\{r(n)\}_{n \in \mathbb{N}}$ , and it has a probability transition matrix:

$$\Theta = \begin{bmatrix} 0 & 0.6 & 0.4 \\ 0.3 & 0 & 0.7 \\ 0.4 & 0.6 & 0 \end{bmatrix},$$

and S-TPMFs were unavailable. Thereafter, the following desired controller gains are derived by utilizing Theorem 3:

$$K_1 = [0.1679, -8.3648, 17.4839], \quad K_2 = [-4.2535, -3.8456, 7.5934], \quad K_3 = [-3.1437, -4.8057, 1.5019].$$

In Figures 2(a) and (b), the electronic throttle is stable with the designed controller, where the S-TPMFs are assumed to satisfy Cases 1 and 2 in Example 1, respectively.

Additionally, let  $\mu_1 = -0.9$ ,  $\mu_2 = -0.2$ , and  $\mu_3 = 0$ . Using Theorem 4, the desired controller gains are derived as follows:

$$K_1 = [0.1679, -8.3648, 17.4839], \quad K_2 = [-4.0598, -5.4513, 10.2426], \quad K_3 = [-6.8307, -4.6026, 5.0295].$$

Figures 2(c) and (d) show that the electronic throttle is stable with the designed controller, where the S-TPMFs are assumed to satisfy Cases 1 and 2 in Example 1, respectively. In addition, according to the values of control gains designed in [49] for the actual experimental platform, the corresponding values in Example 2 are within a reasonable range.

## 5 Conclusion

In this study, the stability problem has been investigated for discrete-time S-MJLSs with unavailable sojourn-time information. A novel analysis method has been developed to study the  $\sigma$ -error mean square stability for discrete-time S-MJLSs, and a set of mode-dependent controllers has been also designed for closed-loop systems, where only the probability transition matrix is used for the semi-Markov chain. Two numerical examples are provided to illustrate the validity of the obtained theoretical results.

In this paper, the stability and stabilization problems have been studied for discrete-time S-MJLSs with unavailable sojourn-time information. In the future, this idea can be expanded in other researches, such as the stability and stabilization of discrete-time (or continuous-time) semi-Markov jump nonlinear systems, and the stability and stabilization of discrete-time hidden semi-Markov jump nonlinear systems.

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